

Fundamentals of Communications

Engineering

Department of Communications Engineering, College of Engineering, University of Diyala, 2016-2017

Class: Second Year

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Room: Comm-02

Lecture: 04

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Convolution Integral

* Finding the effect of a signal on another signal is called convolution.

* The convolution between two signals $f_1(t)$ and $f_2(t)$,

$$f_1(t) \otimes f_2(t) = \int_{-\infty}^{\infty} f_1(\tau) f_2(t-\tau) d\tau \quad \text{--- (70)}$$

* Convolution integral is commutative:

$$f_1(t) \otimes f_2(t) = f_2(t) \otimes f_1(t) \quad \text{--- (71)}$$

* Convolution integral is Distributive

$$f_1(t) \otimes [f_2(t) + f_3(t)] = f_1(t) \otimes f_2(t) + f_1(t) \otimes f_3(t) \quad \text{--- (72)}$$

* Convolution integral is Associative

$$f_1(t) \otimes [f_2(t) \otimes f_3(t)] = [f_1(t) \otimes f_2(t)] \otimes f_3(t) \quad \text{--- (73)}$$

* Convolution can be shifted :-

if $f_1(t) \otimes f_2(t) = g(t)$ then

$$f_1'(t) \otimes f_2(t - t_0) = g(t - t_0) \quad \text{--- (74)}$$

$$f_1(t - t_0) \otimes f_2(t) = g(t - t_0) \quad \text{--- (75)}$$

$$f_1(t - t_0) \otimes f_2(t - t_1) = g(t - t_0 - t_1) \quad \text{--- (76)}$$

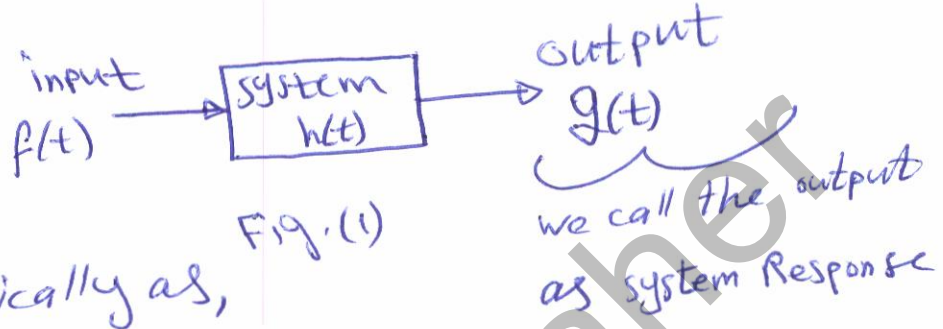
* Convolution with $\delta(t)$:-

$$f(t) \otimes \delta(t) = f(t) \quad \text{--- (77)}$$

Hence :- Convolution of Any
Signal with $\delta(t)$ is the
Signal itself.

System Impulse Response

* In Fig. (1), the system can be described mathematically as,



$$g(t) = f(t) * h(t)$$

(77)

* Thus, to get the effect of any system on a signal, all we need is to do convolution.

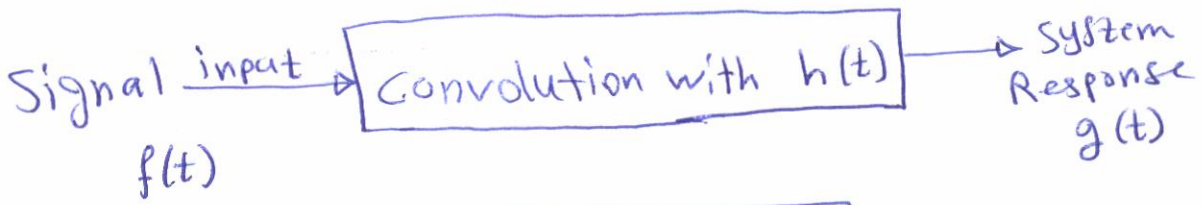
* In other words, Convolution integral is an operation which gives the effect of a system or signal on a system or a signal.



= effect of 1 on 2
effect of 2 on 1



= effect of signal on system / or
effect of system on signal



∴ System response = input \otimes system

* As mentioned previously ∴ convolution of any signal with $\delta(t)$ is the signal itself, then to get the system's function (system transfer function), the input to the unknown system $h(t)$ must be $\delta(t)$.

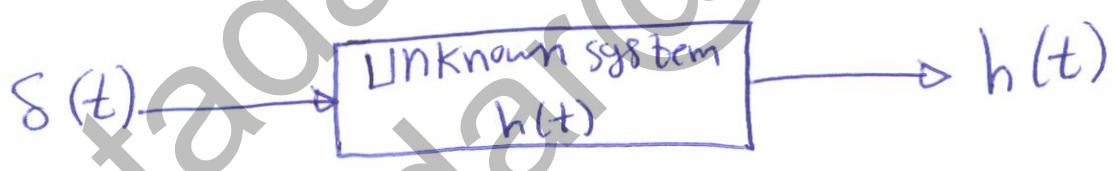
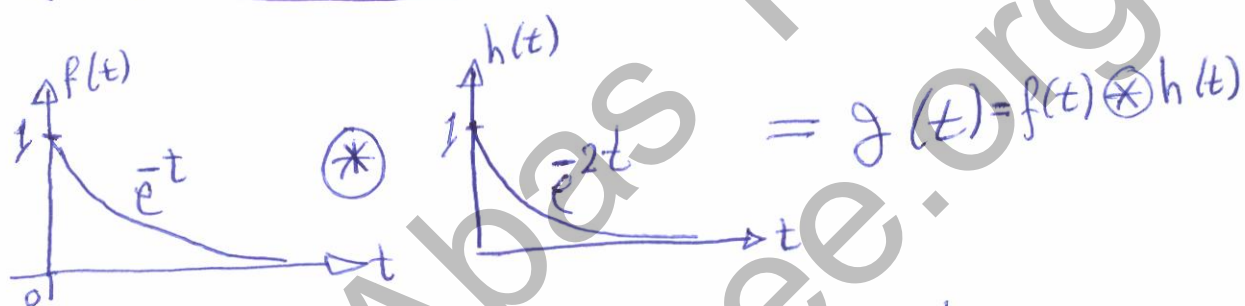
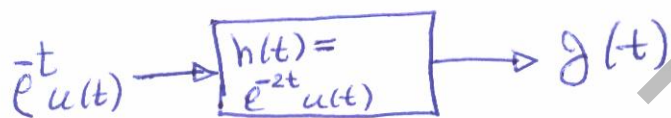


Fig. (2)

* In Fig. (2), the output $h(t)$ is called the system impulse response (which is the transfer function of the system).

EX. For a system with impulse response $h(t) = e^{-2t} u(t)$, determine the response $g(t)$ for the input $e^{-t} u(t)$.

Solution



* both functions starts at zero, $t \geq 0$, then

$$g(t) = f(t) * h(t) = \int_0^t f(\tau) h(t-\tau) d\tau \quad t \geq 0$$

we have $f(\tau) = e^{-\tau} u(\tau)$ & $h(t-\tau) = e^{-2(t-\tau)} u(t-\tau)$

$$\text{then } g(t) = \int_0^t e^{-\tau} e^{-2(t-\tau)} d\tau = \int_0^t e^{-\tau} e^{-2t} e^{2\tau} d\tau \quad t \geq 0$$

$$= e^{-2t} \int_0^t e^{\tau} d\tau = e^{-2t} [e^{\tau}]_0^t = e^{-2t} [e^t - 1] \quad t \geq 0$$

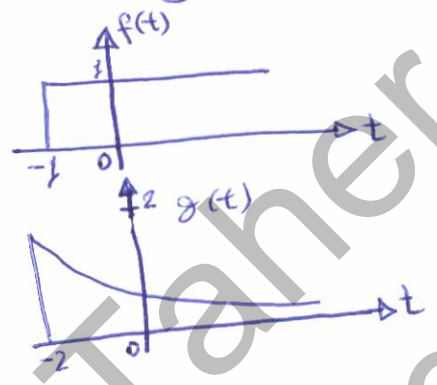
$$= e^{-2t} e^t - e^{-2t} = e^{-t} - e^{-2t} \quad t \geq 0$$

$$g(t) = [e^{-t} - e^{-2t}] u(t)$$

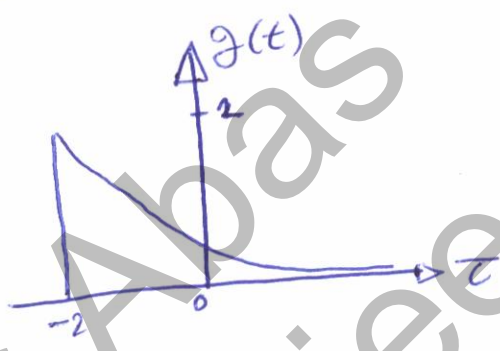
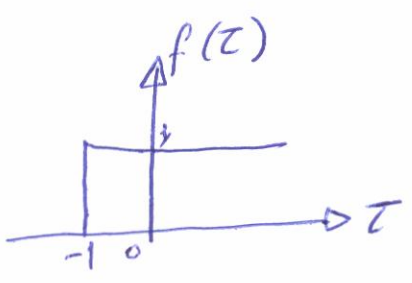
Graphical Convolution

* Lets understand graphical convolution by an example,

Let $f(t) = u(t+1)$
 $g(t) = 2e^{-(t+2)}$
 $c(t) = f(t) \otimes g(t)$

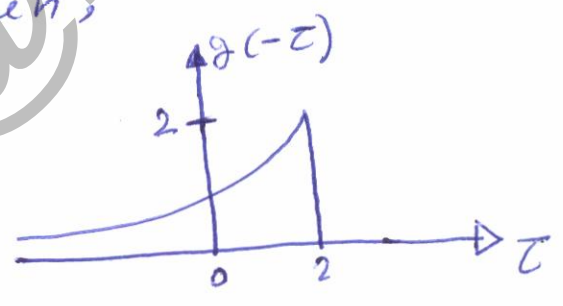


Step 1 replace each t with τ ,



then,

$$c(t) = \int_{-\infty}^{\infty} f(\tau)g(t-\tau) d\tau$$



Step 2 plot $g(-\tau)$

Step 3 shift $g(-\tau)$ by t as follows:

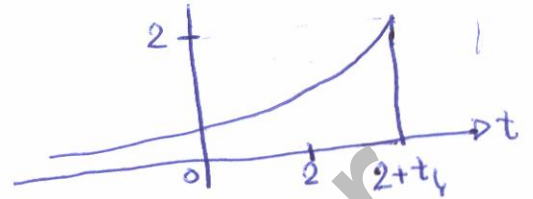
* assume $g(-\tau) = \phi(\tau)$

* now $\phi(\tau)$ shifted by t seconds is $\phi(\tau-t)$

$\phi(\tau-t) = g(-(\tau-t)) = g(t-\tau)$

- * if $+t \rightarrow$ shift to the right \rightarrow
- * if $-t \rightarrow$ shift to the left \leftarrow

Thus, we start to shift to the right by t_1 to get $g(t_1 - \tau)$, $t = t_1 > 0$



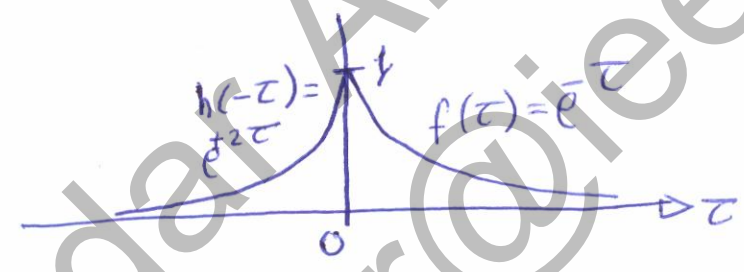
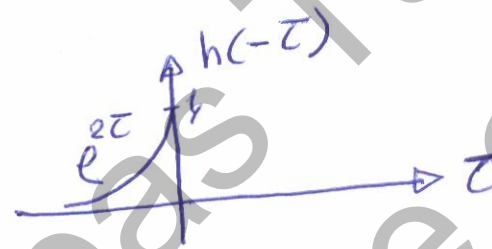
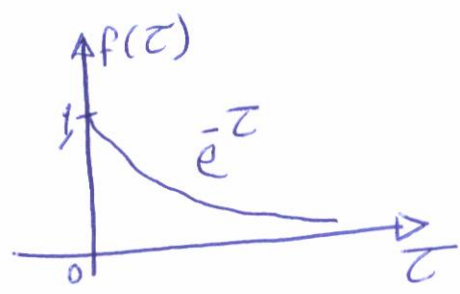
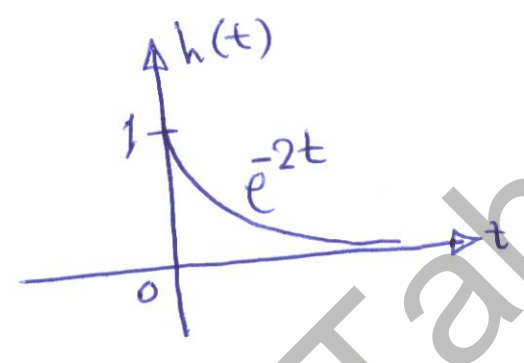
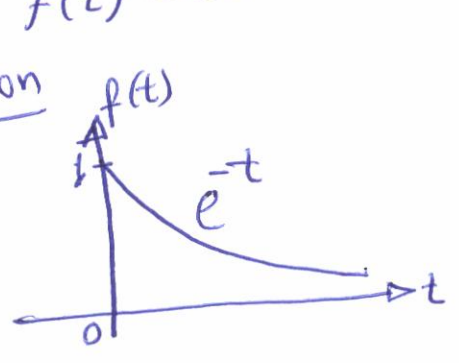
Step 4: Move $g(\tau)$ (scanning) over $f(\tau)$.
- The area under the product of $f(\tau)$ and $g(t_1 - \tau)$ (the shifted frame) is $c(t_1)$, the value of the convolution at $t = t_1$.

Step 5: Repeat this procedure, shifting the frame by different values (positive & negative) to obtain $c(t)$ for all values of t .

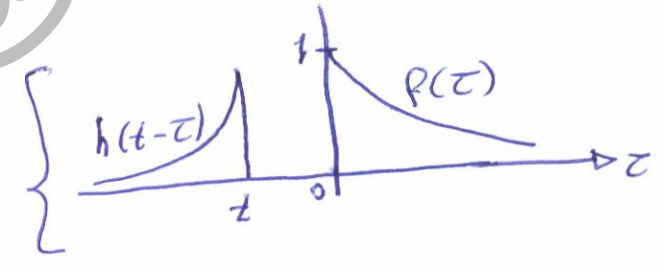
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EX. Determine graphically $y(t) = f(t) \otimes h(t)$ for $f(t) = e^{-t} u(t)$ and $h(t) = e^{-2t} u(t)$.

Solution



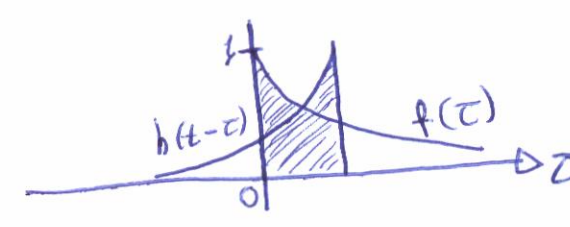
* In this case, there is no overlap:



$$f(\tau)h(t-\tau) = 0 \quad t < 0$$

$$\therefore y(t) = 0, \quad t < 0$$

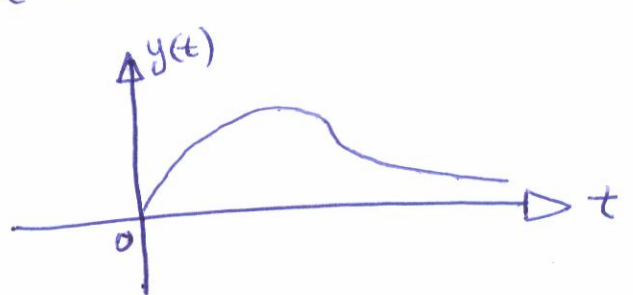
* In this case, there is overlap in the region $0 \leq \tau \leq t$ (shaded area)



$$y(t) = \int_0^t f(\tau) h(t-\tau) d\tau \quad t \geq 0$$

$$y(t) = \int_0^t e^{-\tau} e^{-2(t-\tau)} d\tau = e^{-2t} \int_0^t e^{\tau} d\tau = e^{-t} - e^{-2t} \quad t \geq 0$$

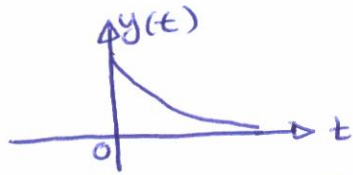
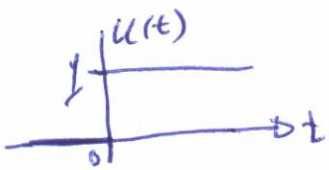
$$\therefore y(t) = (e^{-t} - e^{-2t}) u(t)$$



EX.1 convolve $x(t) = u(t)$ with $y(t) = e^{-t} u(t)$.

(46)

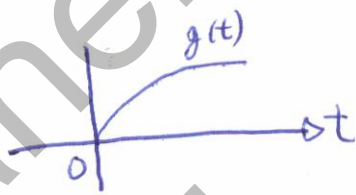
Solution



Let $g(t) = u(t) \otimes e^{-t} u(t) = x(t) \otimes y(t)$

$$g(t) = \int_{-\infty}^{\infty} x(\tau) y(t-\tau) d\tau = \int_0^t e^{-(t-\tau)} d\tau$$

$$= e^{-t} \int_0^t e^{\tau} d\tau = e^{-t} [e^t - 1] = 1 - e^{-t}$$



EX.2 given $r(t) = t \quad t \geq 0$ & $x(t) = e^{-\alpha t} u(t)$, find the convolution $g(t) = r(t) \otimes x(t)$.

Solution

$$g(t) = \int_{-\infty}^{\infty} r(\tau) x(t-\tau) d\tau = \int_0^t \tau e^{-\alpha(t-\tau)} d\tau$$

$$= \int_0^t \tau e^{-\alpha(t-\tau)} d\tau \quad t \geq 0$$

$$= e^{-\alpha t} \int_0^t \tau e^{\alpha \tau} d\tau \quad t \geq 0$$

{ integrate it using }
{ by parts method }

$$= \frac{t}{\alpha} - \frac{1}{\alpha^2} (1 - e^{-\alpha t}) \quad t \geq 0$$

$$g(t) = \left[\frac{t}{\alpha} - \frac{1}{\alpha^2} (1 - e^{-\alpha t}) \right] u(t)$$

EX: convolve $u(t)$ with itself.

Solution

$$y(t) = u(t) \otimes u(t) = \int_{-\infty}^{\infty} u(\tau) u(t-\tau) d\tau$$

$$= \int_0^{\infty} u(\tau) u(t-\tau) d\tau = \int_0^t d\tau = \begin{cases} t & t > 0 \\ 0 & t < 0 \end{cases}$$

OR $y(t) = t u(t)$.

EX: find the effect of the system $h(t) = \frac{1}{t+1} u(t)$ on the input signal $x(t) = u(t)$.

Solution

* input signal $x(t) = u(t)$ (input is D.C.)

$$g(t) = \int_{-\infty}^{\infty} u(t-\tau) \frac{1}{\tau+1} d\tau = \int_0^t \frac{1}{\tau+1} d\tau = \ln(t+1) u(t)$$

EX: How the system $h(t) = u(t)$ will response if its input is $f(t) = u(t) - u(t-2)$?

Solution

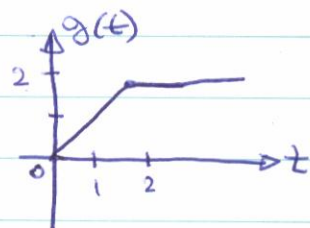
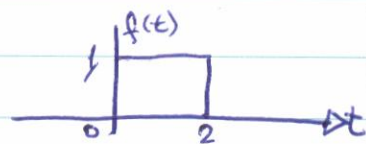
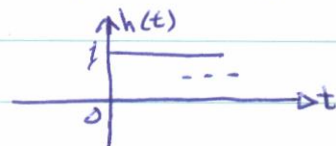
$$g(t) = \int_{-\infty}^{\infty} [u(\tau) - u(\tau-2)] u(t-\tau) d\tau$$

* if $\tau < 0$ and $\tau > 2 \rightarrow g(t) = 0$

$\tau > 0 \rightarrow g(t) = 0$

* if $0 < t < 2 \rightarrow g(t) = \int_0^t d\tau = t$

* if $t > 2 \rightarrow g(t) = \int_0^2 d\tau = 2$



2 Continuous-Time Convolution

The input, $x(t)$, and output, $y(t)$, of a continuous-time LTI system are related by the convolution integral

$$y(t) = \int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau = \int_{-\infty}^{\infty} h(\tau)x(t - \tau)d\tau \quad (8)$$

where $h(t)$ is the impulse response of the system. Consider a system with impulse response and input shown in Fig. 4 and given by

$$h(t) = e^{2t}u(1 - t) \quad (9)$$

$$x(t) = u(t) - 2u(t - 2) + u(t - 5). \quad (10)$$

Substituting these into (8) yields a complicated looking expression,

$$y(t) = \int_{-\infty}^{\infty} [u(t - \tau) - 2u(t - 2 - \tau) + u(t - 5 - \tau)] e^{2\tau}u(1 - \tau)d\tau. \quad (11)$$

This integration problem can be simplified by considering various intervals for t . Flipped and shifted versions of the impulse response are shown in Fig. 5 for these ranges of t . When $t < 1$ the curves only overlap between $t - 5$ and t , see Fig. 5a, so this limits the integration interval. Furthermore, $x(t)$ is a constant value from $t - 5$ to $t - 2$ and a different constant value from $t - 2$ to t , so separating the integral into two terms will greatly simplify our work.

$$\begin{aligned} y(t) &= \int_{t-5}^t [u(t - \tau) - 2u(t - 2 - \tau) + u(t - 5 - \tau)] e^{2\tau}d\tau. \\ &= \int_{t-5}^{t-2} (-1)e^{2\tau}d\tau + \int_{t-2}^t e^{2\tau}d\tau \\ &= -\frac{1}{2}[e^{2(t-2)} - e^{2(t-5)}] + \frac{1}{2}[e^{2t} - e^{2(t-2)}] \\ y(t) &= \frac{1}{2}[1 - 2e^{-4} + e^{-10}]e^{2t} \quad t < 1 \end{aligned} \quad (12)$$

When $t > 1 > t - 2$ (i.e. $1 < t < 3$), the leading edge of $x(t - \tau)$ has shifted out beyond the end of $h(\tau)$, see Fig. 5b. Therefore, the upper limit on the integration becomes the end of $h(\tau)$.

$$\begin{aligned} y(t) &= \int_{t-5}^{t-2} -e^{2\tau}d\tau + \int_{t-2}^1 e^{2\tau}d\tau \\ &= -\frac{1}{2}[e^{2(t-2)} - e^{2(t-5)}] + \frac{1}{2}[e^2 - e^{2(t-2)}] \\ y(t) &= \frac{1}{2}[-2e^{-4} + e^{-10}]e^{2t} + \frac{1}{2}e^2 \quad 1 < t < 3 \end{aligned} \quad (13)$$

When $t - 5 < 1 < t - 2$ (i.e. $3 < t < 6$), the leading section $x(t - \tau)$ has shifted out beyond the end of $h(\tau)$, so the second integral is zero and the upper limit on the first integral is the end of $h(\tau)$, see Fig. 5c.

$$y(t) = \int_{t-5}^1 -e^{2\tau}d\tau = \frac{1}{2}[e^{2(t-5)} - e^2] \quad 3 < t < 6 \quad (14)$$

When $t - 5 > 1$ (i.e. $t > 6$), the curves no longer overlap and

$$y(t) = 0. \quad t > 6 \quad (15)$$

Putting (12) through (15) together yields the final answer,

$$y(t) = \begin{cases} \frac{1}{2} [1 - 2e^{-4} + e^{-10}] e^{2t} & t \leq 1 \\ \frac{1}{2} [-2e^{-4} + e^{-10}] e^{2t} + \frac{1}{2} e^2 & 1 < t \leq 3 \\ \frac{1}{2} [e^{2(t-5)} - e^2] & 3 < t \leq 6 \\ 0 & 6 < t \end{cases} \quad (16)$$

We can use unit step functions to rewrite (16) as

$$\begin{aligned} y(t) &= \frac{1}{2} [1 - 2e^{-4} + e^{-10}] e^{2t} u(1-t) + \left(\frac{1}{2} [-2e^{-4} + e^{-10}] e^{2t} + \frac{1}{2} e^2 \right) [u(3-t) - u(1-t)] \\ &\quad + \frac{1}{2} [e^{2(t-5)} - e^2] [u(6-t) - u(3-t)] \\ &= \left(\frac{1}{2} [1 - 2e^{-4} + e^{-10}] e^{2t} - \frac{1}{2} [-2e^{-4} + e^{-10}] e^{2t} - \frac{1}{2} e^2 \right) u(1-t) \\ &\quad + \left(\frac{1}{2} [-2e^{-4} + e^{-10}] e^{2t} + \frac{1}{2} e^2 - \frac{1}{2} [e^{2(t-5)} - e^2] \right) u(3-t) \\ &\quad + \frac{1}{2} [e^{2(t-5)} - e^2] u(6-t) \\ y(t) &= \frac{1}{2} (e^{2t} - e^2) u(1-t) + (-e^{2t-4} + e^2) u(3-t) + \frac{1}{2} (e^{2t-10} - e^2) u(6-t) \end{aligned} \quad (17)$$

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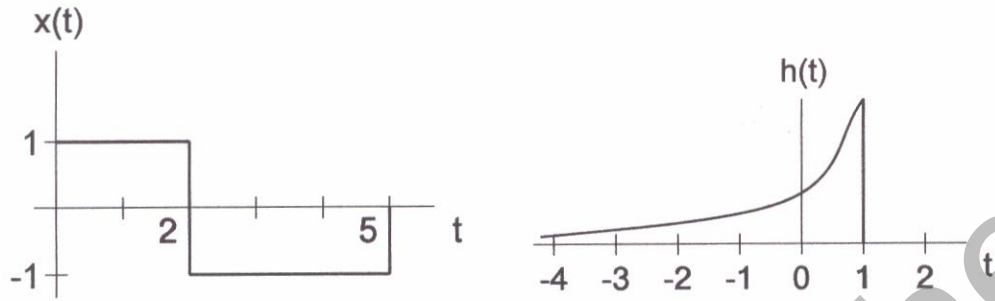


Figure 4: Input signal and system impulse response.

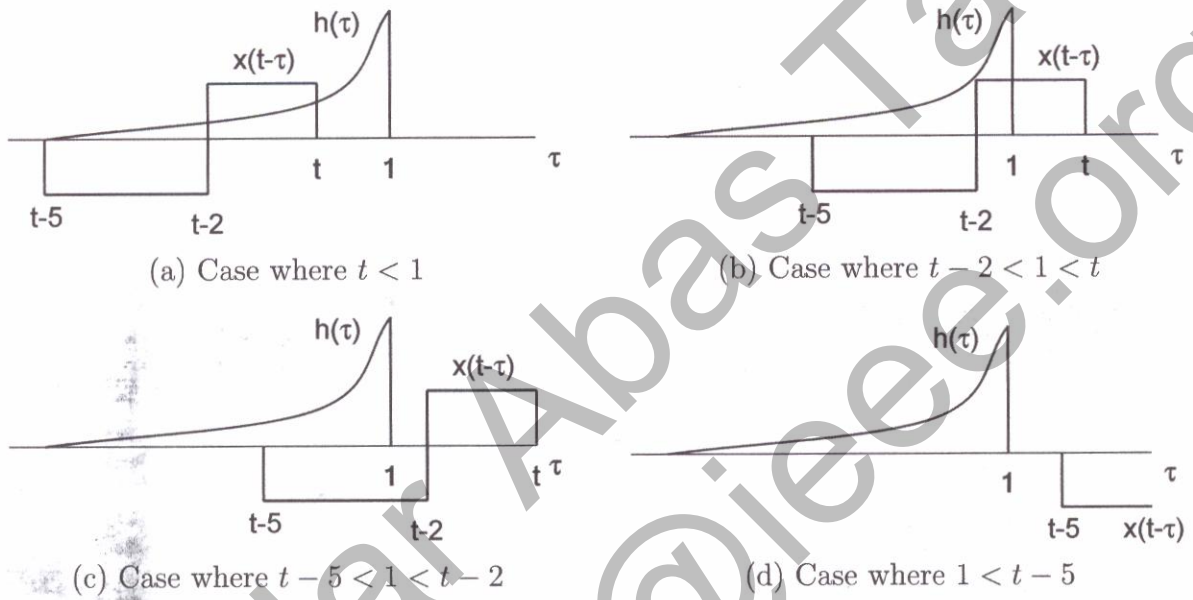


Figure 5: Overlapping curves for various values of t .

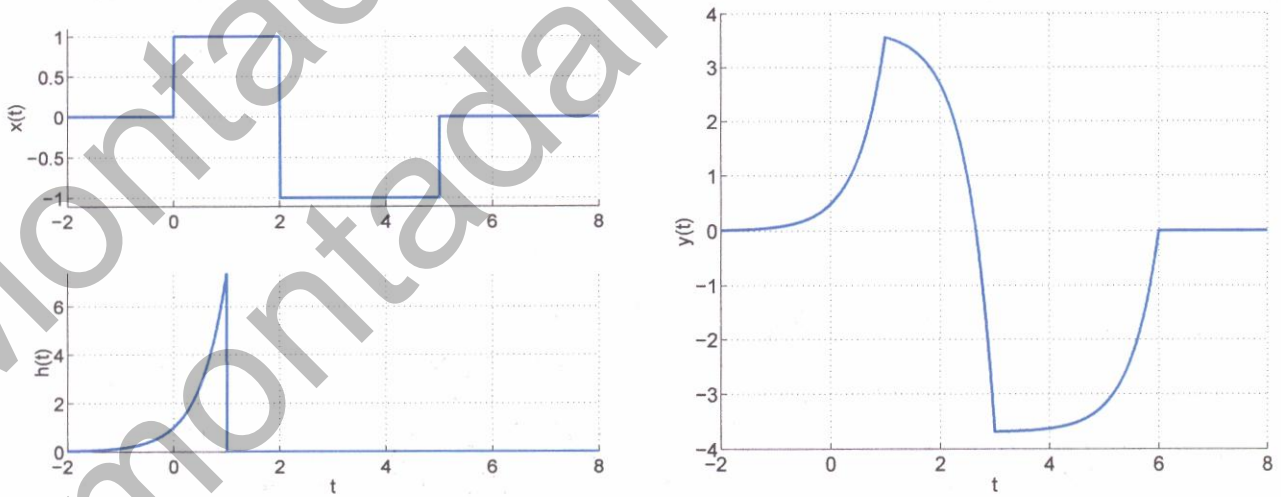


Figure 6: Continuous-time convolution example.